

12. (d) Since  $I_{\text{rms}} = V_{\text{rms}}/Z$  (Equation 23.6), the current is a maximum when the impedance  $Z$  is minimum. The impedance is  $Z = \sqrt{R^2 + (X_L - X_C)^2}$  (Equation 23.7), and it has a minimum value when  $X_C = X_L = 50 \Omega$ .

13. (c) The inductor has a very small reactance at low frequencies and behaves as if it were replaced by a wire with no resistance. Therefore, the circuit behaves as two resistors,  $R_1$  and  $R_2$ , connected in parallel. The inductor has a very large reactance at high frequencies and behaves as if it were cut out of the circuit, leaving a gap in the connecting wires. The circuit behaves as a single resistance  $R_2$  connected across the generator. The situation at low frequency gives rise to the largest possible current, because the effective resistance of the parallel combination is smaller than the resistance  $R_2$ .

14. (a) The capacitor has a very small reactance at high frequencies and behaves as if it were replaced by a wire with no resistance. Therefore, the circuit behaves as two resistors,  $R_1$  and  $R_2$ , connected in parallel. The capacitor has a very large reactance at low frequencies and behaves as if it were cut out of the circuit, leaving a gap in the connecting wires. Therefore, the circuit behaves as a single resistor  $R_1$  connected across the generator. The situation at high frequencies gives rise to the largest possible current, because the effective resistance of the parallel combination is smaller than the resistance  $R_1$ .

15. (e) At low frequencies, the capacitor has a very large reactance. In the series circuit, this large reactance gives rise to a large impedance and, hence, a small current. The parallel circuit has the larger current, because current can flow through the inductor, which has a small reactance at low frequencies.

16. (b) The resonant frequency  $f_0$  is given by  $f_0 = \frac{1}{2\pi\sqrt{LC}}$  (Equation 23.10). It depends only on  $C$  and  $L$ , and not on  $R$ .

17.  $f_0 = 1.3 \times 10^3 \text{ Hz}$

18. (d) The resonant frequency  $f_0$  is given by  $f_0 = \frac{1}{2\pi\sqrt{LC}}$  (Equation 23.10). When a second capacitor is added in parallel, the equivalent capacitance increases (see Section 20.12). Therefore, the resonant frequency decreases.

## CHAPTER 23 | ALTERNATING CURRENT CIRCUITS

### PROBLEMS

**SSM REASONING** The voltage across the capacitor reaches its maximum instantaneous value when the generator voltage reaches its maximum instantaneous value. The maximum value of the capacitor voltage first occurs one-fourth of the way, or one-quarter of a period, through a complete cycle (see the voltage curve in Figure 23.4).

**SOLUTION** The period of the generator is  $T = 1/f = 1/(5.00 \text{ Hz}) = 0.200 \text{ s}$ . Therefore, the least amount of time that passes before the instantaneous voltage across the capacitor reaches its maximum value is  $\frac{1}{4}T = \frac{1}{4}(0.200 \text{ s}) = \boxed{5.00 \times 10^{-2} \text{ s}}$ .

**REASONING** When two capacitors of capacitance  $C$  are connected in parallel, their equivalent capacitance  $C_p$  is given by  $C_p = C + C = 2C$  (Equation 20.18). We will find the capacitive reactance  $X_C$  of the equivalent capacitance from  $X_C = \frac{1}{2\pi f C_p}$  (Equation 23.2).

Because the capacitors are the only devices connected to the generator, the rms voltage  $V_{\text{rms}}$  across the capacitors is equal to the rms voltage of the generator. Therefore, the capacitive reactance  $X_C$  is related to the rms voltage of the generator and the rms current  $I_{\text{rms}}$  in the circuit by  $V_{\text{rms}} = I_{\text{rms}} X_C$  (Equation 23.1).

**SOLUTION** Substituting  $C_p = 2C$  into  $X_C = \frac{1}{2\pi f C_p}$  (Equation 23.2) and solving for  $C$ ,

$$X_C = \frac{1}{2\pi f C_p} = \frac{1}{2\pi f (2C)} = \frac{1}{4\pi f C} \quad \text{or} \quad C = \frac{1}{4\pi f X_C} \quad (1)$$

we obtain

Solving  $V_{\text{rms}} = I_{\text{rms}} X_C$  (Equation 23.1) for  $X_C$  yields  $X_C = \frac{V_{\text{rms}}}{I_{\text{rms}}}$ . Substituting this result

into Equation (1), we obtain

$$C = \frac{1}{4\pi f X_C} = \frac{1}{4\pi f \left( \frac{V_{\text{rms}}}{I_{\text{rms}}} \right)} = \frac{I_{\text{rms}}}{4\pi f V_{\text{rms}}} = \frac{0.16 \text{ A}}{4\pi (610 \text{ Hz})(24 \text{ V})} = \boxed{8.7 \times 10^{-7} \text{ F}}$$

$$I_{1,\text{rms}} = \underbrace{\frac{V_{\text{rms}}}{X_L}}_{L_1 \text{ alone}} = \frac{V_{\text{rms}}}{2\pi f L_1} \quad \text{and} \quad I_{2,\text{rms}} = \underbrace{\frac{V_{\text{rms}}}{X_L}}_{L_2 \text{ alone}} = \frac{V_{\text{rms}}}{2\pi f L_2}$$

The current delivered to  $L_1$  alone is

$$I_{1,\text{rms}} = \frac{V_{\text{rms}}}{2\pi f L_1} = \frac{240 \text{ V}}{2\pi(2200 \text{ Hz})(6.0 \times 10^{-3} \text{ H})} = \boxed{2.9 \text{ A}}$$

The current delivered to the parallel combination of  $L_1$  and  $L_2$  is the sum of that delivered individually to each inductor and is

$$\begin{aligned} I_{\text{P,rms}} &= I_{1,\text{rms}} + I_{2,\text{rms}} = \frac{V_{\text{rms}}}{2\pi f L_1} + \frac{V_{\text{rms}}}{2\pi f L_2} = \frac{V_{\text{rms}}}{2\pi f} \left( \frac{1}{L_1} + \frac{1}{L_2} \right) \\ &= \frac{240 \text{ V}}{2\pi(2200 \text{ Hz})} \left( \frac{1}{6.0 \times 10^{-3} \text{ H}} + \frac{1}{9.0 \times 10^{-3} \text{ H}} \right) = \boxed{4.8 \text{ A}} \end{aligned}$$

13. **SSM WWW REASONING** Since the capacitor and the inductor are connected in parallel, the voltage across each of these elements is the same or  $V_L = V_C$ . Using Equations 23.3 and 23.1, respectively, this becomes  $I_{\text{rms}} X_L = I_{\text{rms}} X_C$ . Since the currents in the inductor and capacitor are equal, this relation simplifies to  $X_L = X_C$ . Therefore, we can find the value of the inductance by equating the expressions (Equations 23.4 and 23.2) for the inductive reactance and the capacitive reactance, and solving for  $L$ .

**SOLUTION** Since  $X_L = X_C$ , we have

$$2\pi f L = \frac{1}{2\pi f C}$$

Therefore, the value of the inductance is

$$L = \frac{1}{4\pi^2 f^2 C} = \frac{1}{4\pi^2 (60.0 \text{ Hz})^2 (40.0 \times 10^{-6} \text{ F})} = 0.176 \text{ H} = \boxed{176 \text{ mH}}$$

14. **REASONING** The inductance  $L$  of the inductor determines its inductive reactance  $X_L$  according to  $X_L = 2\pi f L$  (Equation 23.4), where  $f$  is the frequency of the generator. When the inductor is connected to the terminals of the generator, the rms voltage  $V_{\text{rms}}$  of the generator drives an rms current  $I_{\text{rms}}$  that depends upon the inductive reactance via

# ALTERNATING CURRENT CIRCUITS

$X_L = \frac{V_{\text{rms}}}{I_{\text{rms}}}$  (Equation 23.3). We note that the generator frequency is given in kHz, where  $1 \text{ kHz} = 10^3 \text{ Hz}$ , and the rms current is given in mA, where  $1 \text{ mA} = 10^{-3} \text{ A}$ .

**SOLUTION** Solving  $X_L = 2\pi f L$  (Equation 23.4) for  $L$  yields

$$L = \frac{X_L}{2\pi f} \quad (2)$$

Substituting  $X_L = \frac{V_{\text{rms}}}{I_{\text{rms}}}$  (Equation 23.3) into Equation (1), we find that

$$L = \frac{X_L}{2\pi f} = \frac{\left( \frac{V_{\text{rms}}}{I_{\text{rms}}} \right)}{2\pi f} = \frac{V_{\text{rms}}}{2\pi f I_{\text{rms}}} = \frac{39 \text{ V}}{2\pi(7.5 \times 10^3 \text{ Hz})(42 \times 10^{-3} \text{ A})} = \boxed{0.020 \text{ H}}$$

## REASONING

- a. The inductive reactance  $X_L$  depends on the frequency  $f$  of the current and the inductance  $L$  through the relation  $X_L = 2\pi f L$  (Equation 23.4). This equation can be used directly to find the frequency of the current.
- b. The capacitive reactance  $X_C$  depends on the frequency  $f$  of the current and the capacitance  $C$  through the relation  $X_C = 1/(2\pi f C)$  (Equation 23.2). By setting  $X_C = X_L$  as specified in the problem statement, the capacitance can be found.
- c. Since the inductive reactance is directly proportional to the frequency, tripling the frequency triples the inductive reactance.
- d. The capacitive reactance is inversely proportional to the frequency, so tripling the frequency reduces the capacitive reactance by a factor of one-third.

## SOLUTION

- a. The frequency of the current is

$$f = \frac{X_L}{2\pi L} = \frac{2.10 \times 10^3 \Omega}{2\pi(30.0 \times 10^{-3} \text{ H})} = \boxed{1.11 \times 10^4 \text{ Hz}} \quad (23.4)$$

- b. The capacitance is

**SOLUTION** With only the resistor connected, Equation 20.14 indicates that the resistance is

$$R = \frac{V_{\text{rms}}}{I_{\text{rms}}} = \frac{112 \text{ V}}{0.500 \text{ A}} = 224 \Omega$$

With only the inductor connected, Equation 23.3 indicates that the inductive reactance is

$$X_L = \frac{V_{\text{rms}}}{I_{\text{rms}}} = \frac{112 \text{ V}}{0.400 \text{ A}} = 2.80 \times 10^2 \Omega$$

a. Using these values for  $R$  and  $X_L$  in Equations (1) and (2), we find that the impedance is

$$Z = \sqrt{R^2 + X_L^2} = \sqrt{(224 \Omega)^2 + (2.80 \times 10^2 \Omega)^2} = \boxed{359 \Omega}$$

b. The phase angle  $\phi$  between the current and the voltage of the generator is

$$\tan \phi = \frac{X_L}{R} \quad \text{or} \quad \phi = \tan^{-1} \left( \frac{X_L}{R} \right) = \tan^{-1} \left( \frac{2.80 \times 10^2 \Omega}{224 \Omega} \right) = \boxed{51.3^\circ}$$

19. **SSM REASONING** The voltage supplied by the generator can be found from Equation 23.6,  $V_{\text{rms}} = I_{\text{rms}} Z$ . The value of  $I_{\text{rms}}$  is given in the problem statement, so we must obtain the impedance of the circuit.

**SOLUTION** The impedance of the circuit is, according to Equation 23.7,

$$Z = \sqrt{R^2 + (X_L - X_C)^2} = \sqrt{(275 \Omega)^2 + (648 \Omega - 415 \Omega)^2} = 3.60 \times 10^2 \Omega$$

The rms voltage of the generator is

$$V_{\text{rms}} = I_{\text{rms}} Z = (0.233 \text{ A})(3.60 \times 10^2 \Omega) = \boxed{83.9 \text{ V}}$$

20. **REASONING** The phase angle is given by  $\tan \phi = (X_L - X_C)/R$  (Equation 23.8). When a series circuit contains only a resistor and a capacitor, the inductive reactance  $X_L$  is zero, and the phase angle is negative, signifying that the current leads the voltage of the generator. The impedance is given by Equation 23.7 with  $X_L = 0 \Omega$ , or  $Z = \sqrt{R^2 + X_C^2}$ . Since the

phase angle  $\phi$  and the impedance  $Z$  are given, we can use these relations to find the resistance  $R$  and  $X_C$ .

**SOLUTION** Since the phase angle is negative, we can conclude that only a resistor and a capacitor are present. Using Equations 23.8, then, we have

$$\tan \phi = \frac{-X_C}{R} \quad \text{or} \quad X_C = -R \tan(-75.0^\circ) = 3.73 R$$

According to Equation 23.7, the impedance is

$$Z = 192 \Omega = \sqrt{R^2 + X_C^2}$$

Substituting  $X_C = 3.73 R$  into this expression for  $Z$  gives

$$192 \Omega = \sqrt{R^2 + (3.73 R)^2} = \sqrt{14.9 R^2} \quad \text{or} \quad (192 \Omega)^2 = 14.9 R^2$$

$$R = \sqrt{\frac{(192 \Omega)^2}{14.9}} = \boxed{49.7 \Omega}$$

Using the fact that  $X_C = 3.73 R$ , we obtain

$$X_C = 3.73 (49.7 \Omega) = \boxed{185 \Omega}$$

21. **SSM REASONING** We can use the equations for a series RCL circuit to solve this problem provided that we set  $X_C = 0 \Omega$  since there is no capacitor in the circuit. The current in the circuit can be found from Equation 23.6,  $V_{\text{rms}} = I_{\text{rms}} Z$ , once the impedance of the circuit has been obtained. Equation 23.8,  $\tan \phi = (X_L - X_C)/R$ , can then be used (with  $X_C = 0 \Omega$ ) to find the phase angle between the current and the voltage.

**SOLUTION** The inductive reactance is (Equation 23.4)

$$X_L = 2\pi f L = 2\pi (106 \text{ Hz})(0.200 \text{ H}) = 133 \Omega$$

The impedance of the circuit is

$$Z = \sqrt{R^2 + (X_L - X_C)^2} = \sqrt{R^2 + X_L^2} = \sqrt{(215 \Omega)^2 + (133 \Omega)^2} = 253 \Omega$$

- a. The current through each circuit element is, using Equation 23.6,

$$I_{\text{rms}} = \frac{V_{\text{rms}}}{Z} = \frac{234 \text{ V}}{253 \Omega} = \boxed{0.925 \text{ A}}$$

- b. The phase angle between the current and the voltage is, according to Equation 23.8 (with  $X_C = 0 \Omega$ ),

$$\tan \phi = \frac{X_L - X_C}{R} = \frac{133 \Omega}{215 \Omega} = 0.619 \quad \text{or} \quad \phi = \tan^{-1}(0.619) = \boxed{31.8^\circ}$$

22. **REASONING** The average power dissipated is that dissipated in the resistor and  $\bar{P} = I_{\text{rms}}^2 R$ , according to Equation 20.15b. We are given the current  $I_{\text{rms}}$  but need to find the resistance  $R$ . Since the inductive reactance  $X_L$  is known, we can find the resistance from the impedance, which is  $Z = \sqrt{R^2 + X_L^2}$ , according to Equation 23.7. Since the voltage and the current are known, we can obtain the impedance from Equation 23.6 as  $Z = V_{\text{rms}}/I_{\text{rms}}$ .

**SOLUTION** From Equation 23.7, we can determine the resistance as  $R = \sqrt{Z^2 - X_L^2}$ . With this expression for the resistance, Equation 20.15b for the power becomes

$$\bar{P} = I_{\text{rms}}^2 R = I_{\text{rms}}^2 \sqrt{Z^2 - X_L^2}$$

Using Equation 23.6 to express the impedance, we obtain the following value for the dissipated power

$$\begin{aligned} \bar{P} &= I_{\text{rms}}^2 \sqrt{Z^2 - X_L^2} = I_{\text{rms}}^2 \sqrt{\left( \frac{V_{\text{rms}}}{I_{\text{rms}}} \right)^2 - X_L^2} \\ &= (1.75 \text{ A})^2 \sqrt{\left( \frac{115 \text{ V}}{1.75 \text{ A}} \right)^2 - (52.0 \Omega)^2} = \boxed{123 \text{ W}} \end{aligned}$$

23. **REASONING AND SOLUTION** In order to find the voltage and phase angle we need to know the impedance of the circuit. We know that

$$X_C = \frac{1}{2\pi f C} = \frac{1}{2\pi(4.80 \times 10^3 \text{ Hz})(0.250 \times 10^{-6} \text{ F})} = 133 \Omega$$

The impedance of the circuit can be found using

$$Z = \sqrt{R^2 + (X_L - X_C)^2} = \sqrt{(232 \Omega)^2 + (0 - 133 \Omega)^2} = 267 \Omega$$

The voltage of the generator is

$$V = IZ = (0.0400 \text{ A})(267 \Omega) = \boxed{10.7 \text{ V}}$$

b. The phase angle is

$$\phi = \tan^{-1} \left( \frac{X_L - X_C}{R} \right) = \tan^{-1} \left( \frac{0 - 133 \Omega}{232 \Omega} \right) = \boxed{-29.8^\circ}$$

**REASONING** The rms current  $I_{\text{rms}}$  in the circuit in part c of the drawing is equal to the rms voltage  $V_{\text{rms}}$  divided by the impedance  $Z$  or  $I_{\text{rms}} = \frac{V_{\text{rms}}}{Z}$  (Equation 23.6). The impedance  $Z$  of a series RC circuit is given by Equation 23.7 with  $X_L = 0 \Omega$ , since there is no inductance in the circuit;  $Z = \sqrt{R^2 + X_C^2}$ . The resistance  $R$  is known and the capacitive reactance  $X_C$  can be obtained from the relation  $X_C = 1/(2\pi f C)$  (Equation 23.2), where  $C$  is the capacitance and  $f$  is the frequency. The capacitance, however, is related to the time constant  $\tau$  of the circuit in part a of the drawing. The time constant of an RC circuit is the time for the capacitor to lose 63.2% of its initial charge (see the discussion in Section 20.13), and it is equal to the product of the resistance  $R$  and the capacitance  $C$ ;  $\tau = RC$  (Equation 20.21).

**SOLUTION** Substituting  $Z = \sqrt{R^2 + X_C^2}$  into  $I_{\text{rms}} = \frac{V_{\text{rms}}}{Z}$  gives

$$I_{\text{rms}} = \frac{V_{\text{rms}}}{Z} = \frac{V_{\text{rms}}}{\sqrt{R^2 + X_C^2}} \quad (1)$$

Substituting  $X_C = 1/(2\pi f C)$  (Equation 23.2) into Equation (1) allows us to write the rms current in the circuit in part c of the drawing as follows:

$$I_{\text{rms}} = \frac{V_{\text{rms}}}{\sqrt{R^2 + X_C^2}} = \frac{V_{\text{rms}}}{\sqrt{R^2 + \left( \frac{1}{2\pi f C} \right)^2}}$$

Substituting  $C = \tau/R$  (Equation 20.21) into Equation (1), we arrive at an expression for the rms current: